

Comprehensible Electromagnetics for Comprehensive Engineering

By W. Scott Bennett, Ph.D.

Dr. W. Scott Bennett derives a nanoscale approach to describe the strengths of magnetic and electric near fields that offers an accurate alternative to traditional Hertzian doublet calculations.

Electromagnetic theory has long been based on abstract mathematics; however, to be well-engineered, physical causes and effects must be well understood. That has inspired the cause-based explanations of electromagnetic fields that follow.

An electric dipole is two particles of charge $+q$ and $-q$ separated by a distance $d\ell \rightarrow 0$. And, at any distance $r \gg d\ell$ measured from the midpoint of $d\ell$, the charges $+q$ and $-q$ together cause the electric fields E_{2r} and $E_{2\theta}$ (V/m).



These expressions for E_{2r} and $E_{2\theta}$ result, because $d\ell \rightarrow 0$ and $r \gg d\ell$ for almost all r . And, when $r \gg d\ell$, the propagation distances from $\pm q$ to (r, θ, ϕ) effectively equal $r \mp \cos\theta d\ell/2$. So, the added electric fields of $+q$ and $-q$ can be accurately viewed as

Charged Particle Fields

One microscopic particle with an electric charge of q (Coulombs) in an otherwise empty medium causes the r -directed electric field E_{1r} (Volts/meter).



$$E_{2r} = \frac{r^2 E_{1r}}{\left(r - \frac{\cos\theta d\ell}{2}\right)^2} - \frac{r^2 E_{1r}}{\left(r + \frac{\cos\theta d\ell}{2}\right)^2} = \frac{\left(r + \frac{\cos\theta d\ell}{2}\right)^2 - \left(r - \frac{\cos\theta d\ell}{2}\right)^2}{\left(r^2 - \left(\frac{\cos\theta d\ell}{2}\right)^2\right)^2} r^2 E_{1r} = \frac{2\cos\theta d\ell}{r} E_{1r} \quad (\text{V/m})$$

and

$$E_{2\theta} = \frac{\sin\theta d\ell/2}{r - \frac{\cos\theta d\ell}{2}} E_{1r} + \frac{\sin\theta d\ell/2}{r + \frac{\cos\theta d\ell}{2}} E_{1r} = \frac{\left(r + \frac{\cos\theta d\ell}{2}\right) + \left(r - \frac{\cos\theta d\ell}{2}\right)}{r^2 - \left(\frac{\cos\theta d\ell}{2}\right)^2} \frac{\sin\theta d\ell}{2} = \frac{\sin\theta d\ell}{r} E_{1r} \quad (\text{V/m})$$

because

$$\left(\frac{\cos\theta d\ell}{2}\right)^2 \leq \left(\frac{d\ell}{2}\right)^2 \quad \text{and} \quad \left(\frac{d\ell}{2}\right)^2 \cong 0$$

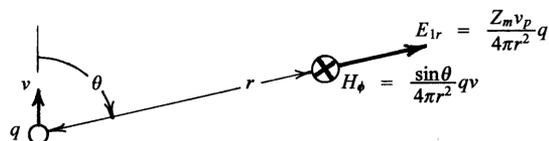
Z_m (Ohms) is the characteristic impedance of the medium containing q ;

v_p (meters/second) is the field propagation velocity in that medium;

r (meters) is the distance from q to the observation point of E_{1r} (V/m); and

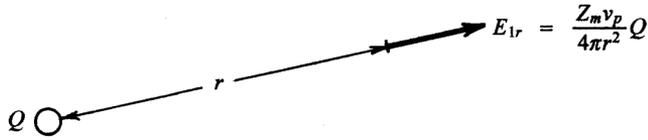
$4\pi r^2$ is an imaginary sphere in which q is centered and on which E_{1r} is observed.

If q moves with a constant velocity of v (meters/second), then qv adds to E_{1r} (V/m) the ϕ -directed magnetic field H_ϕ (Amperes/meter).

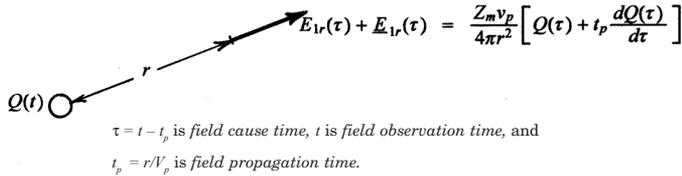


Point Source Fields

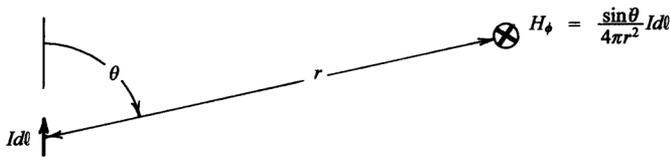
A stationary volume dV containing numerous microscopic particles that have a net electric charge of Q (C) causes the electric field E_{1r} (V/m).



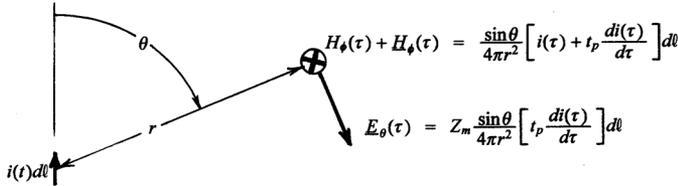
A time-varying net charge of $Q(t)$ (C) in dV will cause the two-component electric field $E_{1r}(\tau) + \underline{E}_{1r}(\tau)$ (V/m).



If some of the particles in dV are moving in the direction of its length dl , a steady movement of net charge, $I dl$ (Axm), causes the magnetic field H_ϕ (A/m).



A time-varying charge movement $i(t)dt$ (Axm) causes the two-component magnetic field $H_\phi(\tau) + \underline{H}_\phi(\tau)$ (A/m) and the electric field $\underline{E}_\theta(\tau) = Z_m H_\phi(\tau)$ (V/m).



A nanocurrent $i(t)dt$ is a volume dV of length dl and cross-sectional area dA , that contains both the moving charge $i(t)dt$ (Axm), and the net charge

$$Q(t) = \int_0^t (i_{in}(t) - i_{out}(t)) dt = dl \int \frac{di(t)}{dl} dt$$

$$= \frac{dl}{v_{pc}} \int di(t) = \frac{i(t)dl}{v_{pc}} \quad (C)$$

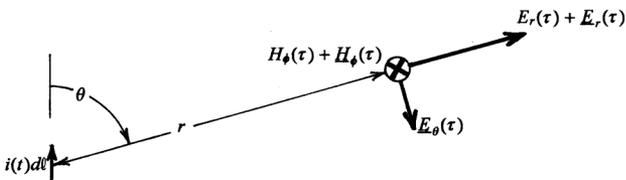
where $v_{pc} = dl/dt$ (m/s) is the field propagation velocity in dV .

Therefore, a nanocurrent $i(t)dt$ has the point-source fields

$$H_\phi(\tau) + \underline{H}_\phi(\tau) = \frac{\sin \theta}{4\pi r^2} [i(\tau) + t_p \frac{di(\tau)}{d\tau}] dl \quad (A/m)$$

$$\underline{E}_\theta(\tau) = Z_m \frac{\sin \theta}{4\pi r^2} [t_p \frac{di(\tau)}{d\tau}] dl \quad (V/m)$$

and
$$E_r(\tau) + \underline{E}_r(\tau) = \frac{Z_m v_p}{4\pi r^2} [i(\tau) + t_p \frac{di(\tau)}{d\tau}] \frac{dl}{v_{pc}} \quad (V/m)$$



The Hertzian Dipole and Its Fields

A Hertzian dipole would be two nanocurrents, one that has end-currents of 0 and $-i(t)$, and the other with end-currents of $i(t)$ and 0. So, an isolated Hertzian dipole would have net charges of

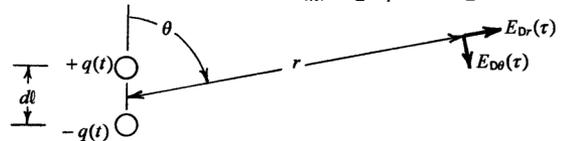
$$\int (0 - i(t)) dt = -q(t) \quad \text{and} \quad \int (i(t) - 0) dt = +q(t) \quad (C)$$

separated by dl , and the fields of $+q(t)$ and $-q(t)$ would add the same as the fields of $+q$ and $-q$ of an electric dipole, to be

$$E_{Dr}(\tau) = \frac{2 \cos \theta dl}{r} [E_{1r}(\tau) + \underline{E}_{1r}(\tau)] = 2 \frac{Z_m \cos \theta}{4\pi r^2} \left[\frac{q(\tau)}{t_p} + i(\tau) \right] dl \quad (V/m)$$

and

$$E_{D\theta}(\tau) = \frac{\sin \theta dl}{r} [E_{1r}(\tau) + \underline{E}_{1r}(\tau)] = \frac{Z_m \sin \theta}{4\pi r^2} \left[\frac{q(\tau)}{t_p} + i(\tau) \right] dl \quad (V/m)$$

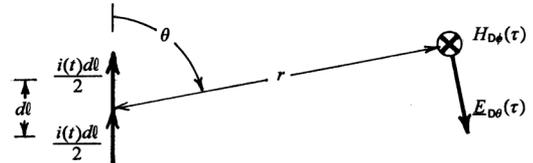


The fields of moving charge would be those of two adjacent nanocurrents, each with an average current of $i(t)/2$ and length dl . Therefore, because $dl \rightarrow 0$, for all $r \gg dl$ the fields of moving charge would equal those of one nanocurrent $i(t)dl$.

$$H_{D\phi}(\tau) = 2 \left[\frac{H_\phi(\tau)}{2} + \frac{\underline{H}_\phi(\tau)}{2} \right] dl = \frac{\sin \theta}{4\pi r^2} \left[i(\tau) + t_p \frac{di(\tau)}{d\tau} \right] dl \quad (A/m)$$

and

$$\underline{E}_{D\theta}(\tau) = \frac{\underline{E}_\theta(\tau)}{2} + \frac{\underline{E}_\theta(\tau)}{2} = \frac{Z_m \sin \theta}{4\pi r^2} \left[t_p \frac{di(\tau)}{d\tau} \right] dl \quad (V/m)$$



Thus, an isolated Hertzian dipole's fields would be

$$E_{Dr}(\tau) = \frac{2Z_m \cos \theta}{4\pi r^2} \left[\frac{q(\tau)}{t_p} + i(\tau) \right] dl \quad (V/m)$$

$$H_{D\phi}(\tau) = \frac{\sin \theta}{4\pi r^2} \left[i(\tau) + t_p \frac{di(\tau)}{d\tau} \right] dl \quad (A/m)$$

and

$$E_{D\theta}(\tau) + \underline{E}_{D\theta}(\tau) = \frac{Z_m \sin \theta}{4\pi r^2} \left[\frac{q(\tau)}{t_p} + i(\tau) + t_p \frac{di(\tau)}{d\tau} \right] dl \quad (V/m)$$

These field equations can be related to textbook equations [1 – 6] for the fields of a Hertzian dipole, as follows. If μ is the permeability of the medium containing $i(t)dl$, and ϵ is the permittivity of that medium, then

$$Z_m = \sqrt{\frac{\mu}{\epsilon}} \quad (\text{Ohms}) \quad \text{and} \quad v_p = \frac{1}{\sqrt{\mu\epsilon}} \quad \left(\frac{\text{meters}}{\text{second}} \right)$$

Electromagnetic Theory

And, if $i(t)$ is assumed to be a sinusoidal current, then as in [1], for example, the replacements

$$i(\tau) \leftarrow I_0 e^{j\omega(t-r/c)}, \quad q(\tau) \leftarrow \frac{I_0 e^{j\omega(t-r/c)}}{j\omega}, \quad \frac{di(\tau)}{d\tau} \leftarrow j\omega I_0 e^{j\omega(t-r/c)}$$

and

$$d\ell \leftarrow L, \quad v_p \leftarrow c, \quad t_p \leftarrow \frac{r}{c}, \quad \text{and} \quad Z_m \leftarrow \frac{1}{\epsilon c}$$

make

$$\begin{aligned} E_{Dr}(\tau) &= Z_m \frac{2 \cos \theta}{4\pi r^2} \left[\frac{q(\tau)}{t_p} + i(\tau) \right] d\ell \\ &= \frac{1}{\epsilon c} \frac{2 \cos \theta}{4\pi r^2} \left[\frac{I_0 e^{j\omega r}}{j\omega r/c} + I_0 e^{j\omega r} \right] L \\ &= \frac{I_0 e^{j\omega(t-r/c)} L \cos \theta}{2\pi \epsilon} \left[\frac{1}{cr^2} + \frac{1}{j\omega r^3} \right] = E_r \end{aligned} \quad (\text{V/m})$$

and

$$\begin{aligned} H_{D\phi}(\tau) &= \frac{\sin \theta}{4\pi r^2} \left[i(\tau) + t_p \frac{di(\tau)}{d\tau} \right] d\ell \\ &= \frac{\sin \theta}{4\pi r^2} \left[I_0 e^{j\omega r} + \frac{r}{c} j\omega I_0 e^{j\omega r} \right] L \\ &= \frac{I_0 e^{j\omega(t-r/c)} L \sin \theta}{4\pi} \left[\frac{j\omega}{cr} + \frac{1}{r^2} \right] = H_\phi \end{aligned} \quad (\text{A/m})$$

and

$$\begin{aligned} E_{D\theta}(\tau) + E_{D\phi}(\tau) &= Z_m \frac{\sin \theta}{4\pi r^2} \left[\frac{q(\tau)}{t_p} + i(\tau) + t_p \frac{di(\tau)}{d\tau} \right] d\ell \\ &= \frac{1}{\epsilon c} \frac{\sin \theta}{4\pi r^2} \left[\frac{I_0 e^{j\omega r}}{j\omega r/c} + I_0 e^{j\omega r} + \frac{r}{c} j\omega I_0 e^{j\omega r} \right] L \\ &= \frac{I_0 e^{j\omega(t-r/c)} L \sin \theta}{4\pi \epsilon} \left[\frac{j\omega}{c^2 r} + \frac{1}{cr^2} + \frac{1}{j\omega r^3} \right] = E_\theta \end{aligned} \quad (\text{V/m})$$

Conclusion

The moving charge Idt of a nanocurrent Idt causes the magnetic field

$$\mathbf{H}_\phi = \frac{\sin \theta}{4\pi r^2} Id\ell \quad (\text{A/m})$$

And, the moving charge Idt also causes the net charge Idt/v_{pc} , which is the cause of the electric field

$$\mathbf{E}_r = \frac{Z_m v_p}{4\pi r^2} \frac{Id\ell}{v_{pc}} \quad (\text{V/m})$$

A time-varying nanocurrent $i(t)dt$ causes the magnetic field

$$(1) \quad \mathbf{H}_\phi(\tau) = \frac{\sin \theta}{4\pi r^2} \left[i(\tau) + t_p \frac{di(\tau)}{d\tau} \right] d\ell \quad (\text{A/m})$$

and the electric field

$$(2) \quad \mathbf{E}_\theta(\tau) = Z_m \frac{\sin \theta}{4\pi r^2} \left[t_p \frac{di(\tau)}{d\tau} \right] d\ell \quad (\text{V/m})$$

Its net charge $i(t)dt/v_{pc}$ causes the electric field

$$\mathbf{E}_r(\tau) = \frac{Z_m k}{4\pi r^2} \left[i(\tau) + t_p \frac{di(\tau)}{d\tau} \right] d\ell \quad (\text{V/m})$$

where $k = v_p / v_{pc}$. And, in very many, if not all, important cases $k=1$, and

$$(3) \quad \mathbf{E}_r(\tau) = \frac{Z_m}{4\pi r^2} \left[i(\tau) + t_p \frac{di(\tau)}{d\tau} \right] d\ell \quad (\text{V/m})$$

The velocities v_p and v_{pc} will be equal when the medium containing $i(t)dt$ is either free space, a vacuum, or air, and the $i(t)$ conductor is copper, lead, aluminum, silver, or gold, for example. That follows, because their relative permeabilities and relative permittivities are all equal to 1.

Based on [1 – 7], on [8], and all of the above – especially on equations (1), (2) and (3) immediately above – it is quite clear that nanocurrents should replace Hertzian dipoles as current elements. The result will undoubtedly be more accurate near-field computations and more comprehensible engineering electromagnetics.

References

1. John D. Kraus, *Antennas*, McGraw-Hill, 1950, pp. 127 – 135.
2. John D. Kraus & Keith R. Carver, *Electromagnetics* (2nd Edition), McGraw-Hill, 1973, p. 616.
3. Richard B. Adler, Lan Jen Chu, and Robert M. Fano, *Electromagnetic Energy Transmission and Radiation*, John Wiley & Sons, 1960, p. 576.
4. William H. Hayt, Jr., *Engineering Electromagnetics* (4th Edition), McGraw-Hill, 1981, p. 482.
5. Simon Ramo, John R. Whinnery & Theodore Van Duzer, *Fields and Waves in Communication Electronics*, John Wiley & Sons, 1965, p. 644.
6. Edward C. Jordan & Keith G. Balmain, *Electromagnetic Waves and Radiating Systems*, Prentice-Hall, 1968, pp. 319 – 320.
7. J.P. Neal, *Electrical Engineering Fundamentals*, McGraw-Hill, 1960.
8. W. Scott Bennett, “Basic Electromagnetics Made True and Clear with Concrete Mathematics,” *High Frequency Electronics*, June 2011, pp. 54 – 61.

About the Author

W. Scott Bennett, Ph.D., served as Assistant Professor at Virginia Polytechnic Institute, where he taught electromagnetics and computer design. He later worked at Hewlett-Packard Company where for 16 years he designed computers and made those designs electromagnetically compatible. Since retiring he has worked to rid basic electromagnetics of abstract mathematics and thus make it easier to understand. He can be reached at: w.scottbennett@juno.com.