

# Design and Optimization of Lumped Element Hybrid Couplers

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This article describes the design, optimization and performance of a hybrid coupler, using accurate lumped element models in the circuit simulation

Quadrature 3 dB hybrid couplers are widely used in microwave circuit design. The principle of a hybrid coupler (Fig. 1) is that the input power is split equally between the coupled and through ports, with a 90 degree phase difference.

The fourth (isolated) port is terminated by a 50-ohm load. Resonant quarter wavelength transmission lines are used in the distributed design approach, and several lumped element design approaches are described in the literature [1, 2]. The design method selected for this example does not use via holes, which makes it easy to implement using either a PCB-based (as in this work) or a MMIC approach. The bandwidth of this design is claimed to be better than the corresponding distributed branch-line hybrid version [1, 2], at least when using idealized L-C models.

In this example, a lumped element coupler was designed with a center frequency of 2.4 GHz using substrate-scalable models for the inductors and capacitors available from Modelithics Inc. The simulation tool used in this work is the Advanced Design System (ADS) from Agilent Technologies. A version of the lumped coupler design was fabricated on a FR-4 substrate with a dielectric constant of 4.3 and measured using an Agilent vector network analyzer.

Simulation comparisons with full-parasitic as well as idealized models for the surface mount components are included to demonstrate the importance of accurate models in software prototyping. The size and performance of the lumped element coupler were

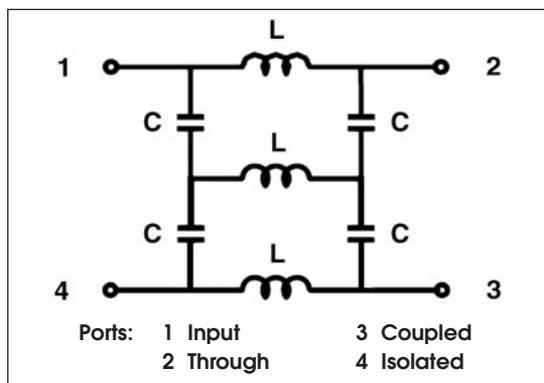


Figure 1 . Basic lumped element hybrid coupler circuit (1).

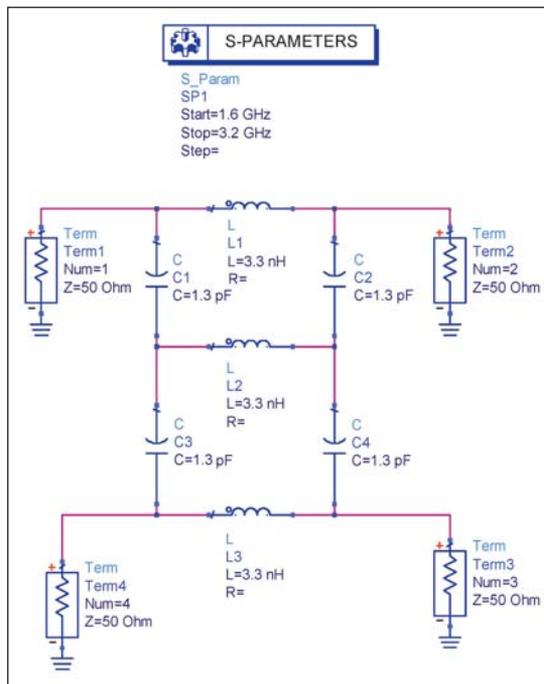


Figure 2 . Ideal lumped hybrid coupler schematic.

HYBRID COUPLER

compared to a distributed branch-line 3 dB coupler designed at the same frequency.

A lumped hybrid coupler such as this may be used in I-Q modulators and demodulators, in balanced amplifiers for VSWR improvement at the input and to increase output power at the output, in single-ended mixer design, and other circuits.

Ideal Design and Simulation

The initial step in the proposed design methodology is to synthesize the nominal values of inductors and capacitors based on the design frequency followed by simulating the ideal design using the ideal lumped element models (Fig. 2). The following equations were used to synthesize the element values based on the frequency and characteristic impedance [1, 2].

$$L = Z_0 / (2\pi f) \text{ nH}$$

$$C = 1 / (Z_0 2\pi f) \text{ pF}$$

The S parameter response and the phase response of the ideal lumped element coupler is as shown in Figures 3 and 4 for:

- Port 1—Input port
- Port 2—Through port
- Port 3—Coupled port
- Port 4—Isolated Port

The second step was to replace the ideal models with Modelithics parasitic models and observe the shift in the parameters of interest ( $S_{11}$ ,  $S_{21}$ ,  $S_{31}$ ,  $S_{41}$  in Figures 5 and 6).

Transmission line effects were added to the previous schematic, using built-in ADS “MLIN,” “MTEE” and “MSTEP” microstrip elements. Figure 7 shows the new schematic; Figure 8 is the constructed coupler. The updated response is as shown in Figures 9 and 10. The response was optimized with a goal of matching the S parameter performance of the ideal design. The L-C part values and the dimensions of the transmission lines

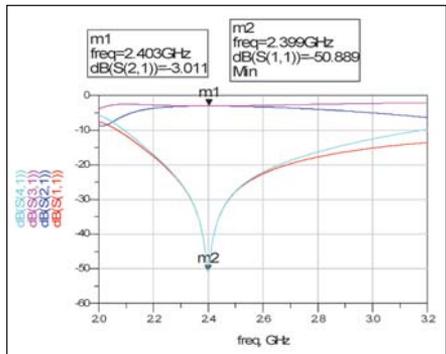


Figure 3 · S parameter response of ideal lumped element coupler.

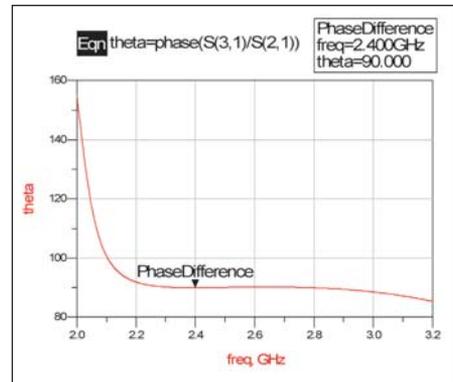


Figure 4 · Phase response of ideal lumped element coupler.

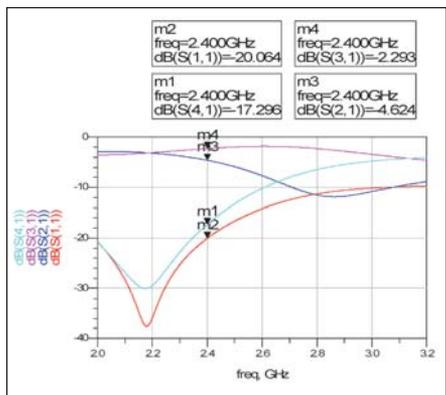


Figure 5 · S parameter response of coupler with Modelithics models.

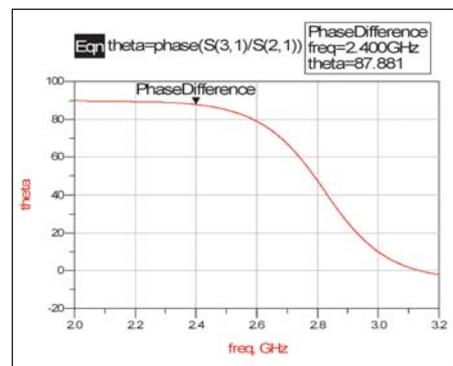


Figure 6 · Phase response of coupler with Modelithics models.

Lumped Elements	Before Optimization (Ideal)	After Optimization (Real)
L1	3.3 nH	1.5 nH
L2	3.3 nH	1.5 nH
L3	3.3 nH	1.5 nH
C1	1.3 pF	3 pF
C2	1.3 pF	3 pF
C3	1.3 pF	3 pF
C4	1.3 pF	3 pF

Table 1 · Inductor and capacitor values before and after optimization.

Dimension	Before Optimization	After Optimization
Wa (mm)	5	5.6
La (mm)	20	7.11
Wb (mm)	5	2.2
Lb (mm)	20	19.9

Table 2 · Dimension of the widths and lengths (MLIN, MSTEP, MTEE) before and after optimization. (Wb, Lb = width, length along vertical; Wa, La = width, length along horizontal).

were allowed to vary, without changing the symmetry of the coupler, and constraining the part values to available ranges.

Table 1 shows component values, before and after optimization. Table 2 shows the widths and lengths of the transmission lines in the layout,

which are optimized without changing the symmetry of the layout, where:

- Wa — Width of the TL along the horizontal of the layout
- La — Length of the TL along the horizontal of the layout

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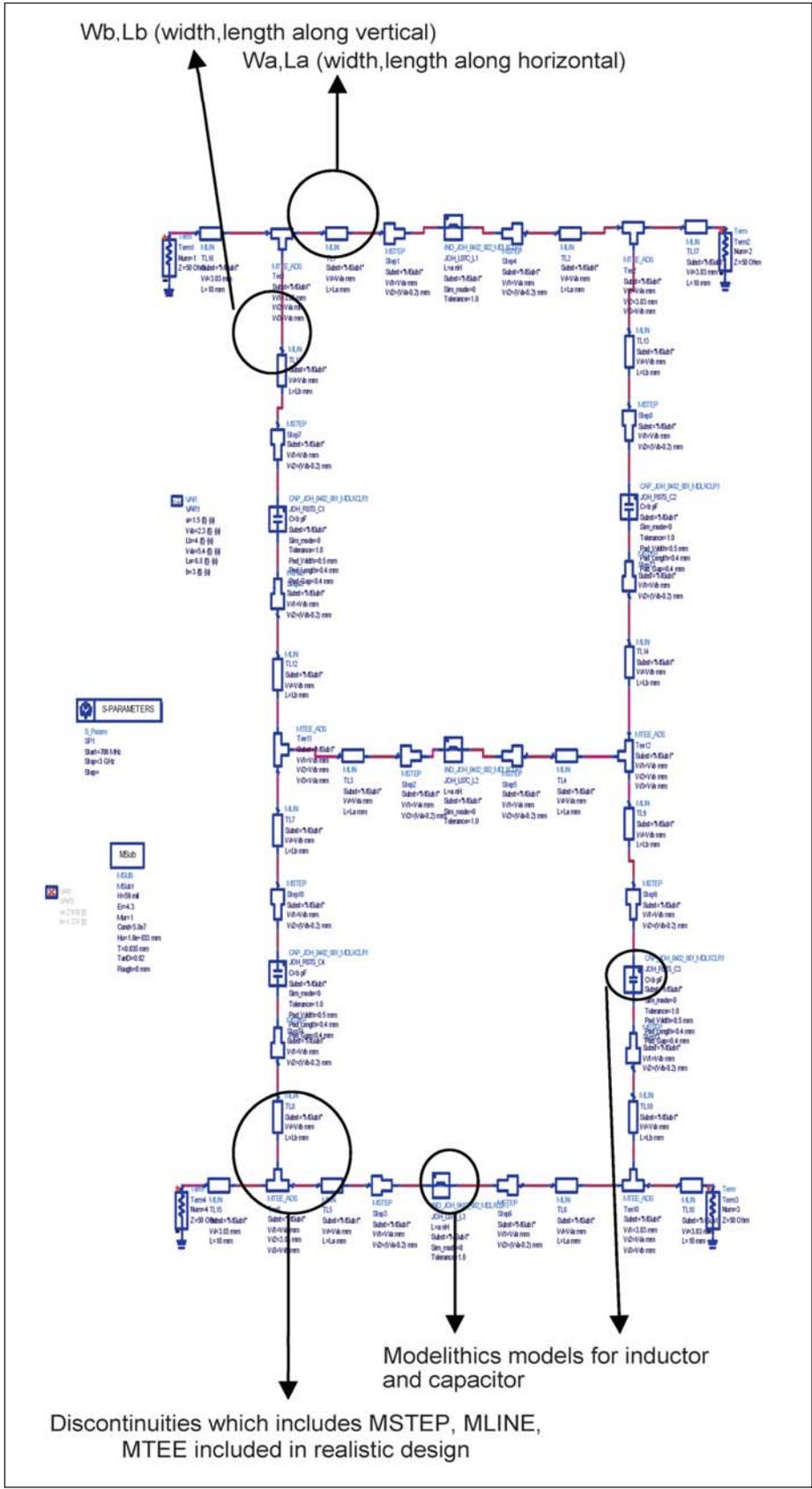


Figure 7 · Schematic of the lumped coupler with transmission line effects.

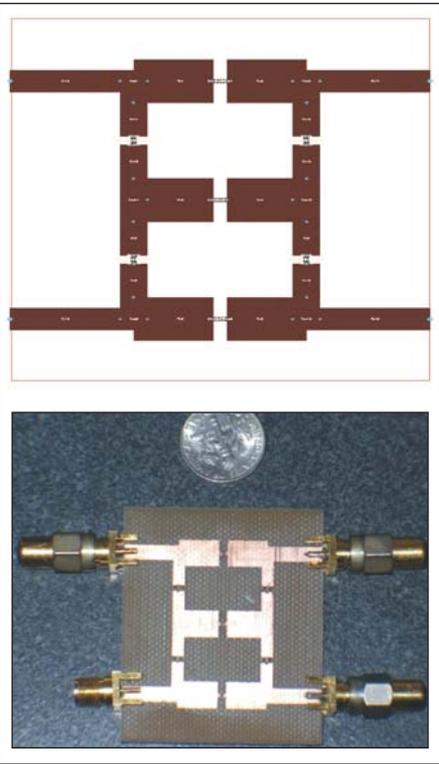


Figure 8 · Layout of the hybrid coupler (top) using ADS and a photo of the constructed unit (bottom).

- Wb — Width of the TL along the vertical of the layout
- Lb — Length of the TL along the vertical of the layout

Layout of the Lumped Coupler

The layout of the lumped coupler was generated using ADS and was used for electromagnetic analysis using Momentum. The simulated electromagnetic results were very close to the circuit simulations using discontinuities.

Distributed Coupler at 2.4 GHz

A branch-line distributed coupler was designed at 2.4 GHz for comparison purposes. The length and width of the transmission lines were calculated using Linecalc for the ideal design. All the discontinuities like MTEE's MSTEP's were included. The simulated response was narrow band when compared to the lumped element response. The response of this

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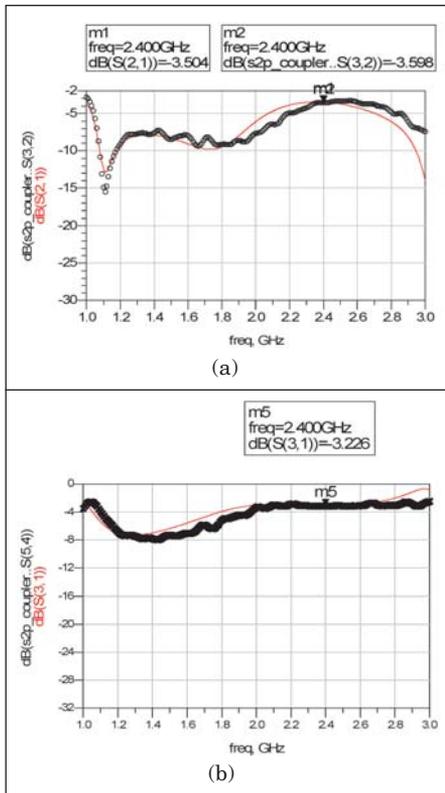


Figure 9 · Measured to modeled results for coupling (a) at port 2 and (b) port 3. Black line is measured; red line is modeled.

distributed version is compared to all the lumped element version

Monte Carlo Tolerance Analysis

A Monte Carlo tolerance analysis was done to analyze the sensitivity of the lumped element coupler to the thickness of the substrate, dielectric constant and the part value tolerance.

The Monte Carlo analysis simulation was run for 30 simulation trials using ADS. The analysis used all the four yield specifications for  $S_{11}$ ,  $S_{21}$ ,  $S_{31}$  and  $S_{41}$ . The data for all the 30 trials were saved and plotted. The expression here refers to return loss to be at least  $-25$  dB. I used three more similar yield specifications for isolation, and insertion loss at coupling ports. Inductor (a), capacitor (b), substrate thickness (Hs), and dielectric constant (Ef) were the vari-

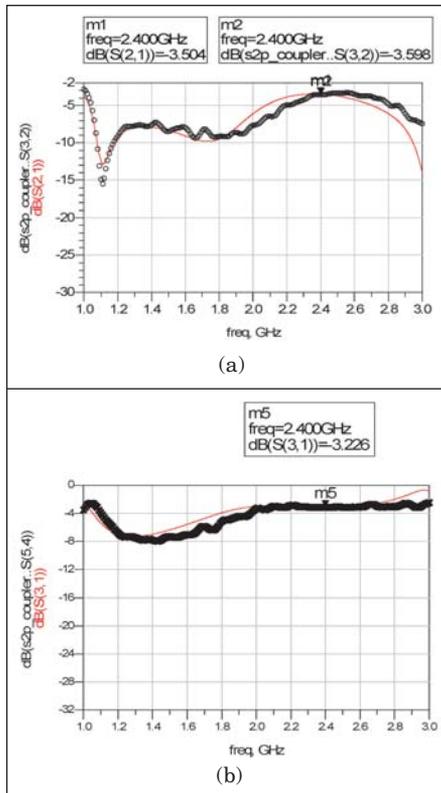


Figure 10 · Measured to modeled results for (a) Return loss and (b) Isolation. Black line is measured; red line is modeled.

ables to which tolerance could be added in the circuit. After reviewing manufacturer data sheet information, the following tolerances were used: 10% for inductors, 5% for capacitors, 5% for substrate thickness and 5%  $t$  for the dielectric constant. The statistical value type form used was Uniform during simulation.

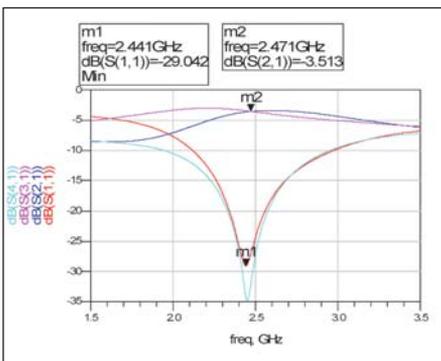


Figure 11 · S parameter response of an ideal distributed coupler.

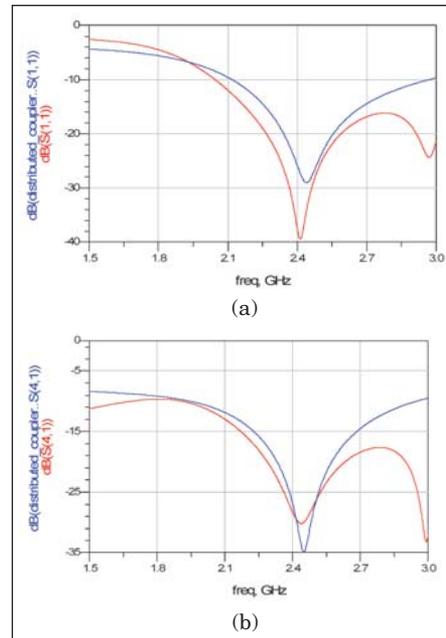


Figure 12 · Performance comparison of lumped coupler with the distributed version at 2.4 GHz (a)  $S_{11}$  (b) Isolation. Blue line is distributed; red line is lumped.

Conclusion

Based on the design literature [1] a hybrid lumped element coupler was designed at 2.4 GHz. Passive accurate models (Modelithics) were used during simulation. The measured to modeled results of the coupler are presented. A Monte Carlo analysis was performed taking into account the board and lumped element tolerances, and the results are presented.

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References

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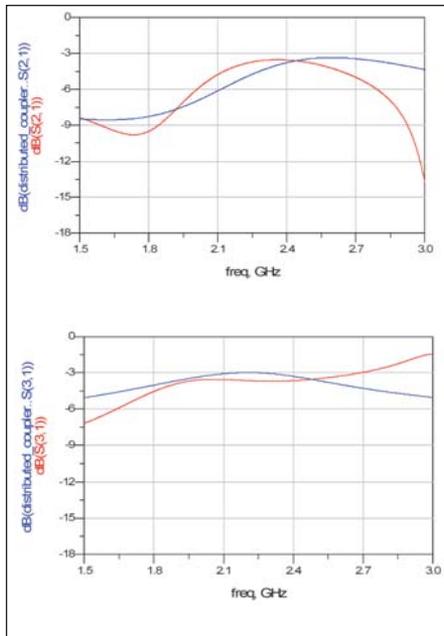


Figure 13 · Performance ( $S$  parameters) comparison of lumped coupler with the distributed version at 2.4 GHz; (a) coupling at port 2 (b) coupling at port 3. Blue line is distributed; red line is lumped.

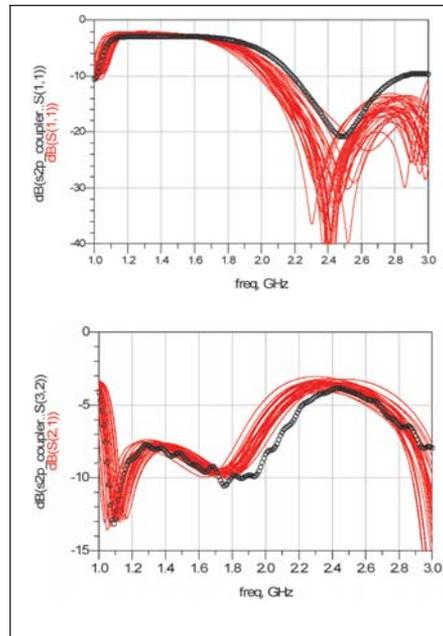


Figure 14 · Tolerance analysis of: (a) return loss and (b) coupling at port 2 (simulated) of the lumped coupler. Black line is measured; red line is modeled.

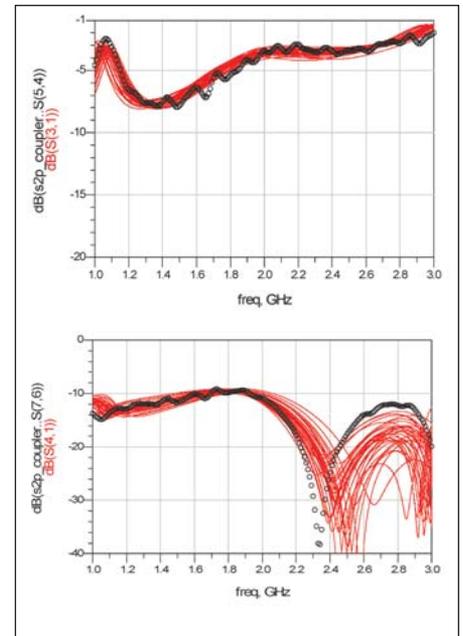


Figure 15 · Tolerance analysis of: (a) coupling at port 3 and (b) isolation (simulated) of the lumped coupler. Black line is measured; red line is modeled.

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